

Data Sheet

Type Description : **Green-Mode PWM Flyback (SSR) Controller**

Product Name : **EST.2700X**

Reversion : **V1.0**

Reversion Date : **May, 2018**

Page : **16 Pages**

EST.2700X Green-Mode PWM Flyback (SSR) Controller



General Description

EST.2700X is a higher integrated PWM flyback controller. It provides several functions to enhance the efficiency and EMI-improved solution, and also built in complete protection.

Meantime, the low startup current, and the proprietary of green-mode function provides gradually mode of frequency reducing under light-load. For zero-load condition, it also built-in burst mode and several parameters to completely turn off PWM output and minimize the power loss of external resistance.

EST.2700X also built-in the leading-edge blanking (LEB) of the current sensing and feedback loop to screen the spike noise from any input signal. The internal slope compensation can limit the constant output over universal AC input range. The sawtooth over frequency function for EMI improved solution.

Meanwhile, EST.2700X also provides various protection, such as, OLP (Over Load Protection), VDD OVP (Over Voltage Protection), Output OVP and VDD OVP to prevent the circuit damage from the abnormal conditions.

EST.2700X is available in SOT-23-6 and SOT-23-8, SOP-8/DIP-8 (with AC pin)

Application

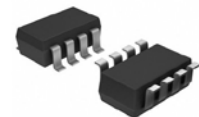
- Switching AC/DC adapter and battery charger
- ATX standby power
- Open frame switching power and CD(R)
- Set-top-boxes(STB) 384Xreplacement

Features

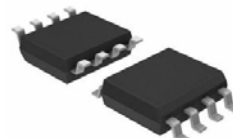
- 100KHz max frequency mode at peak Load
- 65KHz fix frequency mode at PWM Mode
- Adjustable Burst mode
- High voltage CMOS process with excellent ESD protection
- 12ms Soft-start in 65KHz
- Very low startup current (<6 uA)
- 0.5mA ultra-low operating current at light load
- Adaptive Frequency Shuffling and Slope Compensation @ QR and PWM Mode
- Current mode control with Cycle-by-Cycle current limit
- Built-in slope and load regulation compensation
- LEB (Leading-edge blanking) on CS Pin
- UVLO (Under voltage lockout)
- VDD OVP (Over Voltage Protection)
- RTL OVP(Over Voltage Protection) & UVP (Under Voltage Protection)
- OLP (Over load protection)
- Photo coupler short protection
- Feedback open protection



SOT23-6



SOT-23-8



SOP-8L



DIP-8L

Function and Protection Options

Part No.	Package	Freq.	Protection							
		KHZ	VCC OVP	OLP	AUX. OVP	AUX. UVP	CS Open	SDSP	AC-Line OVP	BNO
EST2700A	SOT-23-6L	65KHz	Hiccup	Hiccup / 65mS	Hiccup	Hiccup	Hiccup	Hiccup	NA	NA
EST2700R			Latch	Hiccup / 65mS	Hiccup	Hiccup	Hiccup	Hiccup		
EST2700L			Latch	Latch / 65mS	Latch	Latch	Hiccup	Hiccup		
EST2700M		100KHz	Hiccup	Hiccup / 42mS	Hiccup	Hiccup	Hiccup	Hiccup	NA	NA
EST2700H		135KHz	Hiccup	Hiccup / 32mS	Hiccup	Hiccup	Hiccup	Hiccup	NA	NA
EST2700B	SOT-23-8 SOP-8L DIP-8L	65KHz	Hiccup	Hiccup / 65mS	Hiccup	Hiccup	Hiccup	Hiccup	Hiccup	Hiccup

Note: EST lead-free products contain molding compounds/die attach materials and 100% matte tin plate termination finish; which are fully compliant with RoHS. EST lead-free products meet or exceed the lead-free requirements of IPC/JEDEC J-STD-020C for MSL classification at lead-free peak reflow temperature. EST defines "Green" to mean lead-free (RoHS compliant) and halogen free (Br or Cl does not exceed 900ppm by weight in homogeneous material and total of Br and Cl does not exceed 1500ppm by weight)

EST.2700X Green-Mode PWM Flyback (SSR) Controller



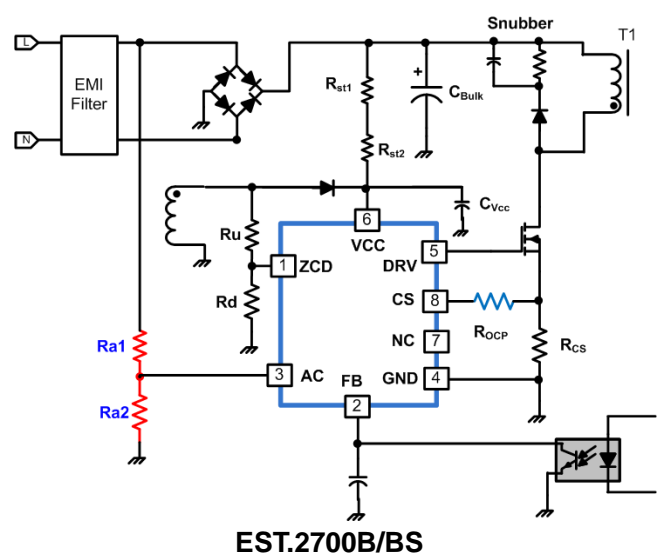
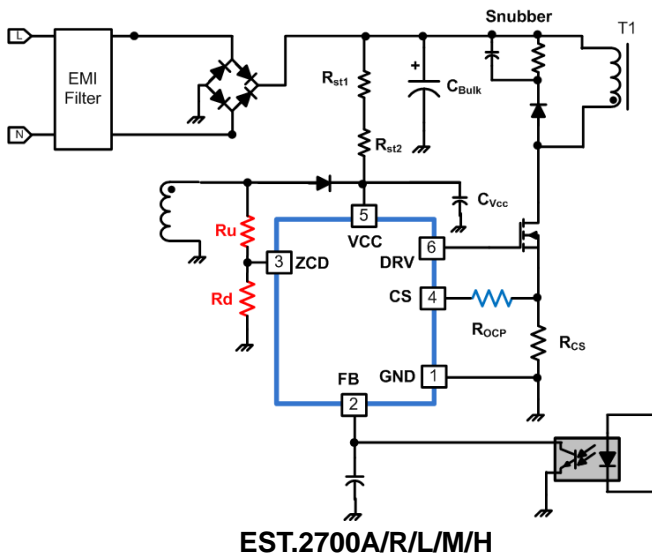
Ordering Information

Part Number	Package	Packaging	Note
EST2700A/R/L/M/H	SOT-23-6	Tape & Reel	Green
EST2700BS	SOT-23-8	Tape & Reel	Green
EST2700BS	SOP-8L	Tape & Reel	Green
EST2700B	DIP-8L	Tape	Green

Pin Assignments and Package Type

SOT-23-8L SOP-8L DIP-8L	SOT-23-6L	NAME Description	Description
1	3	ZCD	Auxiliary voltage sense, Quasi Resonant and Vout OVP&UVP detection.
2	2	FB	Voltage input pin by connecting a photo-coupler
3	--	AC	AC Brown in/out and Line OVP detection.
4	1	GND	Ground
5	6	DRV	Driver output to driver the external MOSFET
6	5	VDD	Power supply pin
7	--	NC	
8	4	CS	Current Sense input. The current sense resistor between this pin and GND is used for current limit setting.

Application Circuit



EST.2700X Green-Mode PWM Flyback (SSR) Controller



Absolute Maximum Ratings

Parameter Symbol	Symbol	Limit Values		Unit	Remark
		Min.	Max		
Supply Voltage V_{DD}	V_{DD}	-0.3	32	V	
AC,FB,CS,ZCD Voltage	$V_{AC}, V_{FB}, V_{CS}, V_{ZCD}$	-0.3	7	V	
Gate Driver Voltage	V_{GATE}	-0.3	$V_{DD}+0.3$	V	
Operation Junction Temperature	T_j	-40	125	°C	
Operation Ambient Temperature	T_A	-25	85	°C	
Storage Temperature	T_{stg}	-55	150	°C	
Power Dissipation	PD	-	408	mW	SOT23-6
			556		SOP-8
Junction-to-Ambient Thermal Resistance*	θ_{JA}	-	245	°C/W	SOT23-6
			180		SOP-8
Junction-to-Case Thermal Resistance**	θ_{JC}	-	55	°C/W	SOT23-6
			39		SOP-8
Lead temperature (Soldering, 10 sec)		-	260	°C	
ESD Voltage Protection	HBM	$V_{ESD-HBM}$	-	3.0	KV
	MM	V_{ESD-MM}	-	300	V

Stress beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

*Free standing with no heatsink; without copper clad.(Measurement condition – just before junction temperature T_J enters into OTP)

**Measure on the DRAIN pin close to plastic interface

***Measure on the PKG top surface

Recommended Operating Conditions

Parameter Symbol	Symbol	Limit Values		Unit	Remarks
		Min.	Max		
Supply Voltage V_{DD}	V_{DD}	9	26	V	
Startup Resistor Value	R_{star}	1	14	MΩ	
Ambient temperature range	T_{opr}	-40	85	°C	

DC Electrical Characteristics ($V_{CC} = 15V, T_a = 25^\circ C$)

Supply Voltage (VCC Pin):

Parameter	Symbol	Min.	Typ.	Max.	Unit	Conditions
Startup Current	I_{CC-ST}	2	3.5	5.5	μA	UVLO ON - 0.1V
Operating Current (with 1nF load on DRV pin)	I_{CC-OP}	0.4	0.6	0.8	mA	$V_{FB}=0V$
	I_{CC-OP}	1.5	2	2.5	mA	$V_{FB}=2.5V$ CL=1nF
	I_{CC-OLP}	0.2	0.35	0.5	mA	OLP
UVLO (off)	$V_{UVLO-OFF}$	7	7.5	8	V	
UVLO (on)	$V_{UVLO-ON}$	15	16	17	V	
V_{DD} OVP Level	V_{OVP}	27	28	29	V	
OVP Debounce Time	T_{OVP}		100		uS	V_{DD} sweep until no OUT pulse

Voltage Feedback(FB Pin):

Parameter	Symbol	Min.	Typ.	Max.	Unit	Conditions
Short Circuit Current	I_{Zero}	0.10	0.14	0.18	mA	$V_{FB}=0V$
Open Loop Voltage	V_{FB-OP}	4.8	5.0	5.2	V	FB pin open
Burst mode start voltage(on)	V_{BUR_ON}	0.9	1.0	1.1	V	
Burst Mode Hysteresis	V_{BUR_HY}	0.05	0.1	0.15	V	
OLP Trip Level		3.8	4.0	4.2	V	
Delay Time of FB pin Open loop protection	t_{Delay_OLP}	14		16	ms	

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Current Sensing (CS Pin):

Parameter	Symbol	Min.	Typ.	Max.	Unit	Conditions
Leading Edge Blanking Time	T_{LEB}	200	300	400	ns	
Propagation Delay to Output	T_{pd}	80	100	120	nS	Guarantee by Design
Maximum CS Off Voltage1	V_{CSTH1}	0.55	0.6	0.65	V	$I_{AUX} \geq 150\mu A$ **
Maximum CS Off Voltage2	V_{CSTH2}	0.65	0.7	0.75	V	$I_{AUX} < 150\mu A$ **
OCP source current	I_{OCP}	18.5		21.5	%	I_{OCP}/I_{AUX}
Over Load Protection	V_{OLP}	0.45	0.5	0.55	V	
Debounce Time of OLP	T_{OLP}	54	64	74	mS	
Short Circuit Protection Voltage	V_{SCP}	0.8	0.9	1.0	V	
Debounce Time of V_{SCP}	T_{SCP}		4		cycle	Guarantee by Design

Zero Crossing Detector(ZCD Pin):

Parameter	Symbol	Min.	Typ.	Max.	Unit	Conditions
Output OVP Trigger Point	V_{TH_OVP}	2.9	3	3.1	V	
Output OVP Deglitch Time Constant	T_{OVP_delay}		4		Cycle	Guarantee by Design
Output UVP Trigger Point	V_{TH_OVP}	0.7	0.8	0.9	V	
Output UVP Deglitch Time Constant	T_{OVP_delay}		4		Cycle	Guarantee by Design
Positive Clamped voltage	V_{POS}	6		7	V	
Negative Clamped voltage	V_{NEG}	-0.05		0.05	V	
High Low line boundary current	I_{Line}	130	150	170	μA	

Alternating Current Detect(AC Pin):

Parameter	Symbol	Min.	Typ.	Max.	Unit	Conditions
Brown In trigger point	V_{BNI}	0.8	0.85	0.9	V	
Brown Out trigger point	V_{BNO}	0.7	0.75	0.8	V	
BNO De-bounce time	T_{BNO}	110	125	140	mS	
Line OVP trigger point	V_{LNOVP}	2.95	3	3.05	V	
Line OVP release point	V_{LNOVP_HYS}	2.85	2.9	2.95	V	
Line OVP De-bounce time	T_{LNOVP}	110	125	140	mS	
AC detect time	T_{ACD}	21.5	22	22.5	mS	

Driver(DRV Pin) :

Parameter	Symbol	Min.	Typ.	Max.	Unit	Conditions
Output Low Level	V_{OL}			1	V	$V_{DD} = 16V, I_O = 20mA$
Output High Level	V_{OH}	8			V	$V_{DD} = 16V, I_O = 20mA$
Output Clamp Voltage Level	V_{G_Clamp}	11	12.5	14	V	$V_{DD} = 25V$
Rising Time	T_R	200	300	400	nS	$V_{DD} = 16V, C_L = 1nF$
Falling Time	T_F	10	30	50	nS	$V_{DD} = 16V, C_L = 1nF$

Timer Section:

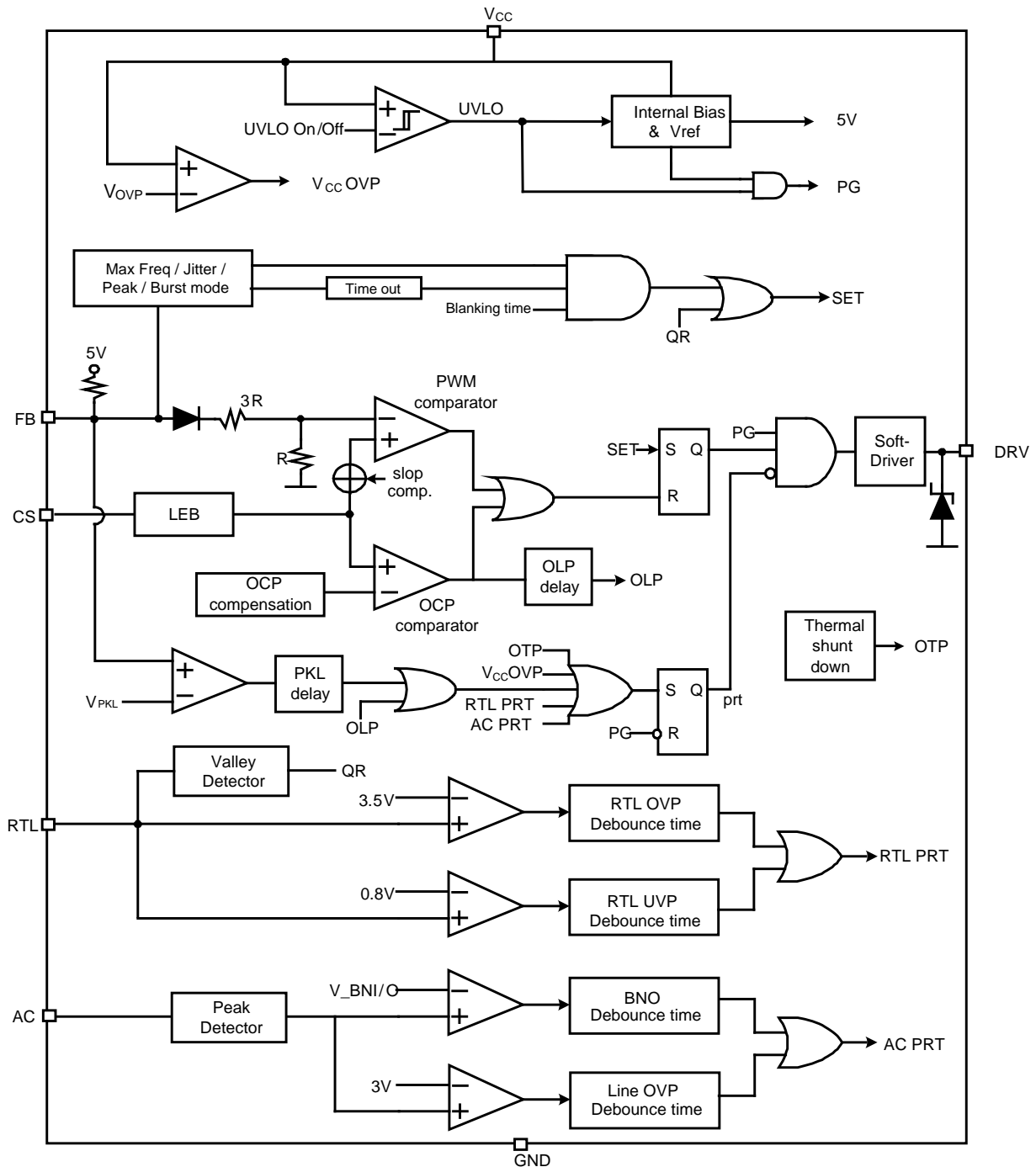
Parameter	Symbol	Min.	Typ.	Max.	Unit	Conditions
Burst Mode Frequency	F_{Burst}	20	22.5	25	KHz	
PWM Mode Frequency	F_{PWM}	61	65	69	KHz	
Peak Mode Frequency	F_{Peak}	120	130	140	KHz	
Voltage stability of Frequency	F_{PSRR}	-1		+1	%	$V_{DD} = 11V \sim 25V$
Frequency Shuffling Range	F_{jitter}	+/-4	+/-6	+/-8	%	
Maximum duty cycle	D_{MAX}	70	75	80	%	
Internal Soft Startup Time	T_{SS}	10		15	mS	

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Block Diagram EST2700X



Application Note

Operation Overview

The EST.2700X meets the green power requirement and very suitable for the use in those networking adaptors, TV open frame and various consumer power, which can provide more power efficiency and lower power loss. It also supports various kind of protection for every abnormal environments.

V_{DD} Start-up and Control

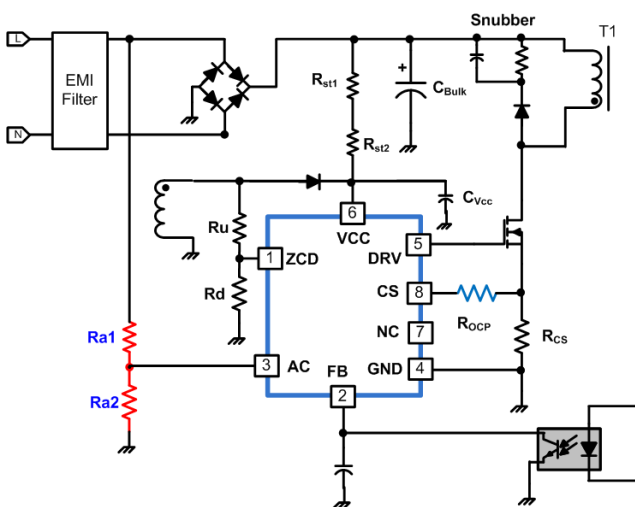


Fig-1.

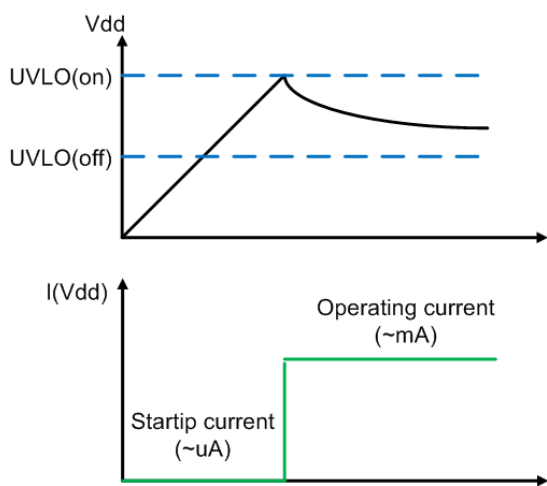


Fig-2.

The start-up circuit of EST.2700X is shown in Fig.1 . It's internal comparator will detect the voltage on the Vcc pin, and assures the supply voltage enough to turn on the EST.2700X. At beginning, the startup current is provided by (R_{st1}/R_{st2}) to charge the capacitor C_{VCC} till V_{CC} get enough voltage (UVLO_ON) to turn on itself,

refers to fig.2. Meantime, it go a step further to deliver the gate drive signal to enable the Aux. winding of transformer , and then provides supply current. The startup current of EST.2700X is designed to be very low so that C_{VCC} could be charged up above UVLO_on threshold level and it starts up quickly.

EST.2700X series is process with low power mix-mode process (5V and 32V), which max start-up current is below 5.5uA. R-start calculate as below :

$$\frac{V_{bulk} - V_{UVLO_ON}}{R_{start}} > I_{cc-st}$$

It is trade off between startup time and a higher startup resistance. Therefore, carefully select the value of Rstart, C_{VCC} to optimize the power consumption and startup time.

SS, Soft-start Sequence

EST.2700X also built-up 12.5ms (typical) soft-start to soften the electrical stress occurring in the power supply during startup, refer to Fig.3. As soon as V_{CC} reaches UVLO_on, the Cs peak voltage is gradually increased from 0.2V to the maximum level, see fig.4.

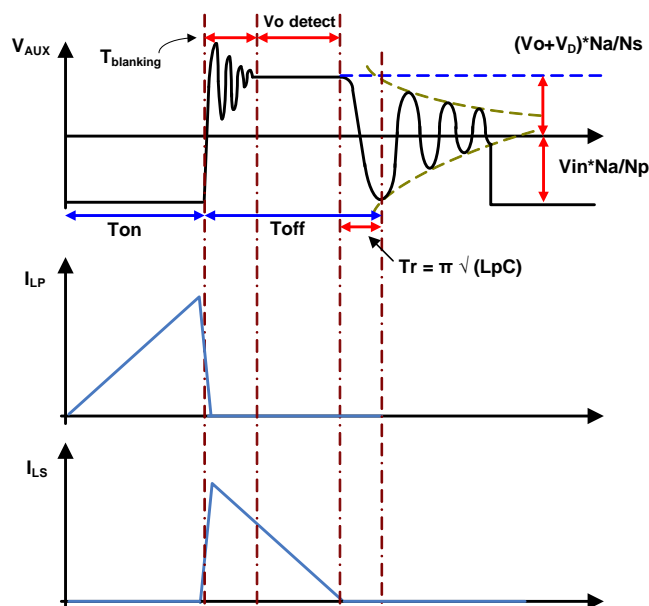


Fig.3

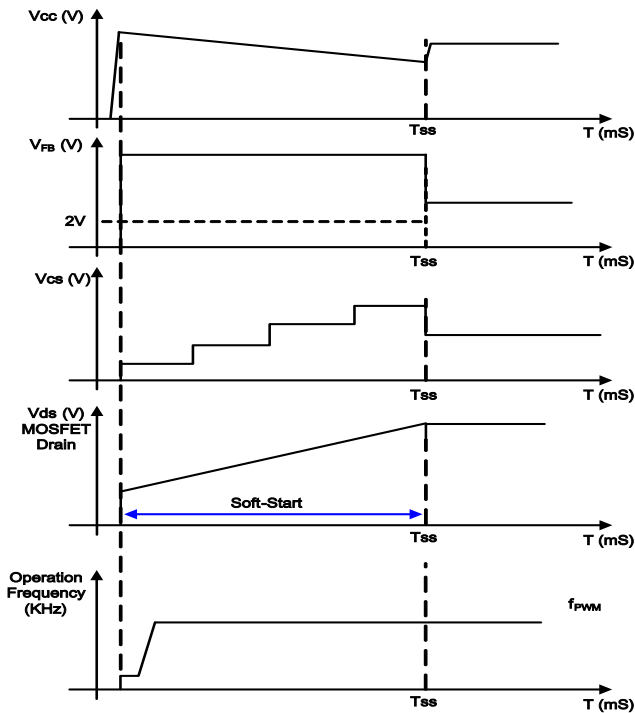


Fig.4

V_{CC} Stimulation Mode

EST.2700X provides stimulation mode to avoid abnormal re-start-up under heavy loading to no-load, caused by non-balance of discharge of V_{CC} cap and output cap, which is different with burst mode. The waveform is shown in fig.5.

Condition : V_{FB} < V_{BUR_ON} & V_{CC} < 9.5V trigger, Hysteresis Voltage 1V

Action : IC fix output F_{Burst}, and V_{CS} keeps as 0.15V

Notice : Design V_{AUX} higher than 11V

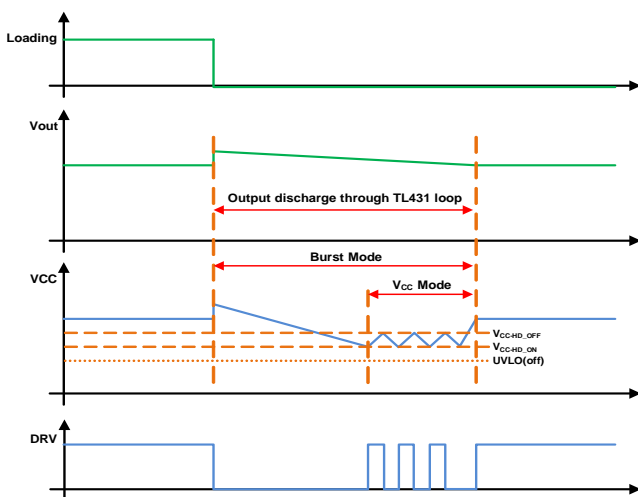


Fig.5

ZCD: Demagnetization Detection from ZCD pin (QR Mode Detection)

After MOSFET turns off, the current of secondary side diodes goes down to zero, and then the transformer core will be demagnetized completely, see fig.5. At the same time, a quasi resonant signal will be detected from auxiliary winding by ZCD pin through the external resistor divider.

Programmable V_{O_OVP} & burst mode level

This ZCD pin is also used to program the burst level at light load and high output voltage at system open loop. A resistive divider between Aux winding and GND is used to set a voltage at this pin to determine the peak current level when power enters the adaptive burst mode. At the same time, it also detects voltage level of output.

V_{O_OVP} : Calculate the ratio of R_d to (R_u+R_d)

$$\frac{R_d}{R_u + R_d} = \frac{V_{TH_OVP}}{(V_{O_OVP} + V_d)} \times \frac{N_s}{N_a}$$

Adjust Burst :

$$V_{BUR_ON} = (V_o + V_d) \times \frac{N_a}{N_s} \times \frac{R_d}{R_u + R_d} - 1.7$$

FB, Voltage Feedback Loop

EST.2700X series adopt current mode control, that is say, the voltage feedback signal is provided from TL431 at secondary side through the photo-coupler to FB pin and compare to the current signal sensing from Cs pin at primary side of MOS current to control the on/off of MOSFET.

In order to enhance light load efficiency, the loss of the feedback resistor in parallel with photo-coupler is reduced. Due to small feedback resistor current, shunt regulator selection and minimum regulation current design have to be considered more carefully to make sure it's able to regulate under low cathode current.

To make sure the stability of feedback is very important. Unstable feedback signal will introduce output oscillate or

audio noise. You can monitor the ripple & Noise of output to adjust the phase and gain margin of close loop.

- (1). R_{bias1} and R_{bias2} to prevent the abnormal output voltage at heavy loading. Generally, we suggest R_{bias1} 100~1K Ω , R_{bias2} 1.5~2.5K Ω ;
- (2). R_{phase}/C_{phase} is for RC phase compensation and prevent oscillate to adjust the value of C_{FB}
- (3). The ratio of R_3 and R_{3A} is depend on the output voltage spec (TL431 , $V= 2.5V$)

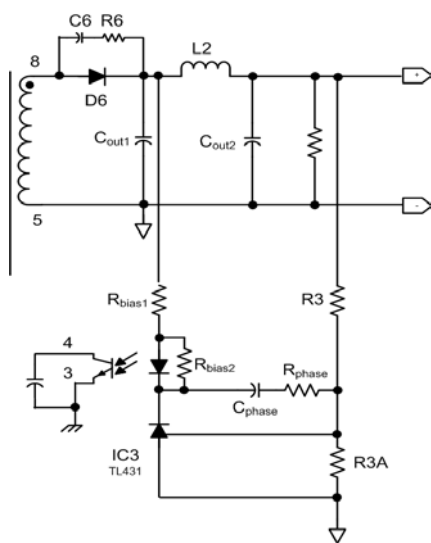


Fig.6

In addition, V_{FB} is also used to determine the green mode level .When V_{FB} is under V_{BUR_ON} , it is under no load or light load condition; at light loading, burst mode can effectively reduces the switching loss. When V_{FB} is larger than V_{BUR_ON} , it will leave away the standby mode. The normal operation of V_{FB} is from V_{BUR_ON} to 2.4V; meanwhile, short-circuit current is around I_{Zero} .

CS, Current sense Loop

Current mode PWM control mode detects the current command (CS) from the Rsense (the primary MOSFET current sense resistor) and voltage command from photocoupler (FB) to determine whether the system reaches a stable or not. There is a potential risk of sub-harmonic when the duty of flyback methodology is larger than 50% and the operation under continues conduction mode (CCM) , therefore, EST.2700X series of

built-in high and low slope compensation to avoid the sub-harmonic risk.

A leading-edge blanking (LEB) time is included in the input of CS pin to prevent the false-trigger from the current spike. Meanwhile, it is strongly recommended to put a suitable R-C filter for higher power application to prevent the CS pin being damaged by the unknow negative spike.

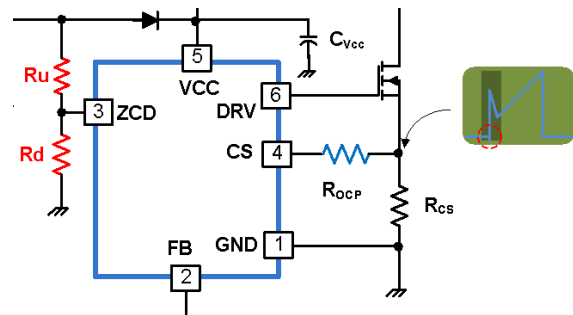


Fig-7.

DRV

The driving capability of EST.2700X is around 450mA, which can support power rate around 60~70W, and it is limited the maximum duty-cycle below 80% to avoid the transformer saturation.

Typically, the threshold of MOSFET is about 20V, and the maximum clamp voltage of EST.2700X is 14V to prevent breakdown of MOSFET.

Complete Protection

EST.2700X integrates various kind of protection to make sure operation safety.

VCC OVP (Over Voltage Protection)

The maximum ratings of the EST.2700X are around 32V. To prevent the V_{CC} enter breakdown condition, EST.2700X series are integrated with OVP function on V_{CC} pin. Whenever the V_{CC} voltage is higher than the V_{OVP} threshold, the output gate drive circuit will be turn-off simultaneously and the power MOSFET is turn-off until the next UVLO(on) cycle.

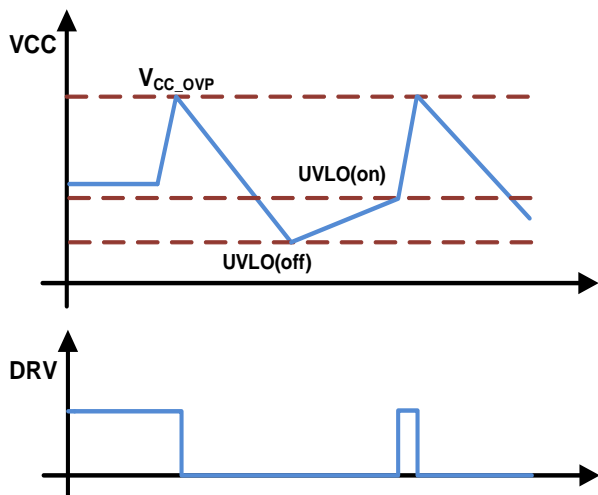


Fig.8

SCP (Short Circuit Protection)

A resistive divider between Aux winding and GND is used to monitor output voltage. When output circuit is short,

Therefore, as V_{ZCD} is lower than 0.8V during date off region, then V_{TH_UVP} is triggered, EST.2700X is to enable UVP function in order to reduce input power

OLP (Over Load Protection)

EST.2700X has new OLP built-in at CS pin, and its merit of close loop methodology makes audio noise free.

The adjustment of OCP is through ZCD and CS, please refer to Fig.7. It can detect the status of AC line and output voltage through the resistance divider (R_u, R_d) by the reflection waveform of Aux-winding. At negative cycle, V_{ZCD} will keep "0" and output I_{OCP} at CS pin to change the level of slope compensation, please see Fig.7. Therefore, it can modify the R_u and R_{OCP} to get target of OCP @full range.

Step 1. Short R_{OCP} and modifies R_{CS} to target of OCP@90Vac

Step 2. Increase R_{OCP} impedance to reduce OCP and check the OCP of AC full range. Modifies R_{OCP} to make OCP unanimous for AC full range.

SDSP, Secondary Diode Short Protection

After short circuit of 2nd side schottky, the

inductance current is too low to discharge completely caused by lower output voltage, and then it will continues to increasing to induce abnormal saturation of transformer during LEB timing, therefore, higher peak current induce serious high V_{ds} to damage MOSFET. EST.2700X detects the inductance current through the resistance, R_{cs} , of CS pin, and will trigger protection (latch or hiccup) when V_{cs} higher than 0.85V and sustains 3cycle timing.

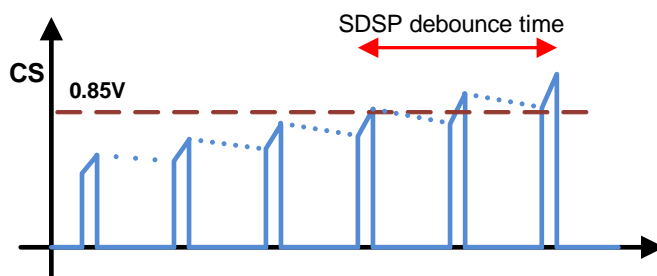


Fig.9

AC, Brown-in/out & Line OVP

IN2P083B provides real detection of AC line through AC pin connected directly to AC line. When the V_{cc} of IN2P083B reaches $UVLO_ON$, it is into the state of AC detection, and sustains a delay time T_{ADC} .

This AC pin is used to program the AV over and under voltage level through a resistive divider (R_{a1}/R_{a2}).

If V_{ac} is lower than below V_{BNI} or higher than V_{LNOVP_HYS} , It will turn-off the output till next cycle to check the condition is removed or not. Even after it turn-off, this pin is continues to detect line status. If V_{ac} is lower than below V_{BNO} or higher than V_{LNOVP} for the timing T_{BNO} and T_{LNOVP} , it will be turn-off, and re-start again. Please refers to fig10.

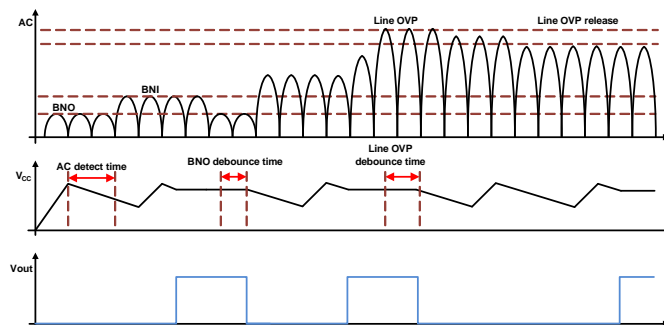


Fig.10

EST.2700X Green-Mode PWM Flyback (SSR) Controller



Layout Guide line

Better layout plan can reduce unknow noise no matter signal or EMI, please refer to the list below :

- ✓ **Big current path** : A&B (Area 1) area are high frequency current loop, line to line is as close as possible, and avoid near low voltage control area
- ✓ **Low voltage area** : R divider need to be as near FB_Pin as possible.
- ✓ **Secondary Side Schottky** : routing as close as possible

Grounding : (2).(3) and (4) grounding separated with each other, and end connects to (1) ground.

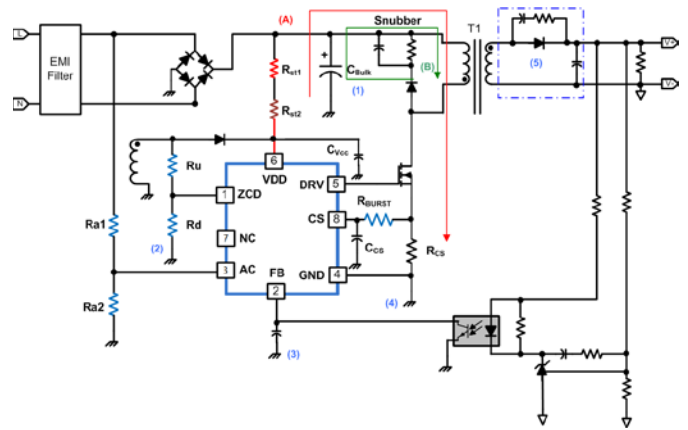


Fig.11

Table 1: Complete Protection

Issue	Protection	Pin	Protection Conditions	
1st	V-Sense	VCC OVP	Vcc	$V_{cc} > 28V$
1st	V-Sense	VCC UVLO Off	Vcc	$V_{cc} < 7.5V$
1st	V-Sense	Brown In Fail	AC	$V_{AC} < 0.85$
1st	V-Sense	Brown out	AC	$V_{AC} > 0.75V$
1st	V-Sense	Line OVP	AC	$V_{AC} > 3.0V$
1st	V-Sense	T1 Aux gnd open	ZCD	ZCD UVP trigger
1st	V-Sense	MOS short/Gate to GND	ZCD	ZCD UVP trigger
1st	V-Sense	CS pin open	CS	$V_{CS} > 0.7V$ after 4 cycles
1st	ZCD	ZCD upper R open	ZCD	ZCD UVP : after soft-start $ZCD < 0.85V$ & $FB > 4V$
1st	ZCD	ZCD upper R short	ZCD	ZCD OVP : $ZCD > 3V$ & $FB > 4V$
1st	ZCD	ZCD down side open	ZCD	ZCD OVP : $ZCD > 3V$ & $FB > 4V$
1st	ZCD	ZCD down-side short	ZCD	ZCD UVP : after soft-start time $ZCD < 0.85V$ & $FB > 4V$
2nd	SDSP	2nd side Schottkey short	CS	$V_{CS} > 0.85V$ after 4 cycles
2nd	SCP	Output short	ZCD	1. 12ms blank time during start-up 2. after 4 cycles 3. ZCD UVP = $0.8V$ & $FB > 4V$ trigger
2nd	OVP	Output OVP	ZCD	V_{ZCD} compares to $3V$ through the resistance divider
1nd	OCP	OCP	CS	Hi/Low line OCP external adjust , Max current limit $CS = 0.7V$
2nd	OLP	OLP	CS	$CS > 0.5V$
2nd	Short before power on		ZCD	1. 12ms blank time at start-up 2. after 4 cycles 3. ZCD UVP = $0.8V$ & $FB > 4V$ trigger
2nd	Short after power on		ZCD	1. after 4 cycles 2. ZCD UVP = $0.8V$ & $FB > 4V$ trigger
IC	Chip OTP			chip OTP at $150\text{ }^{\circ}C$

EST.2700X Green-Mode PWM Flyback (SSR) Controller



EST.2700X Green-Mode PWM Flyback (SSR) Controller

Package Information

SOT-23-6L:

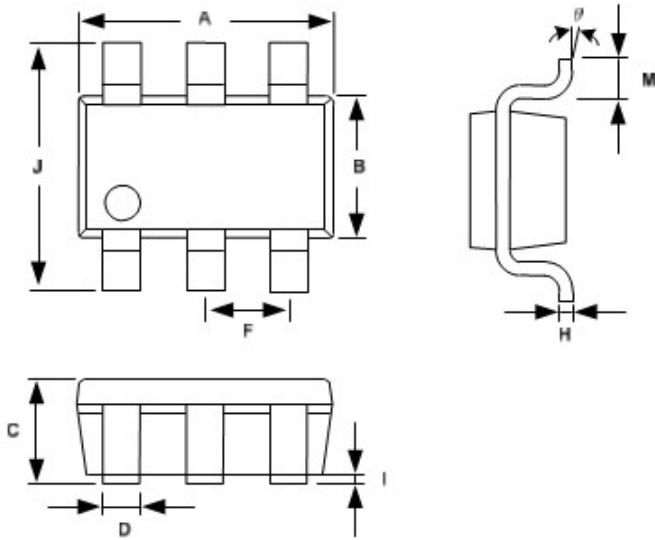
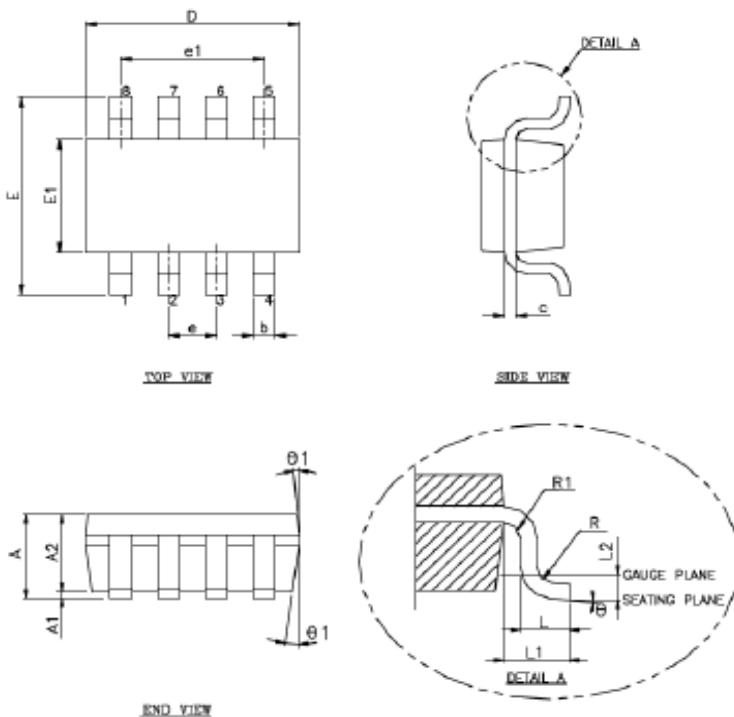


Fig 8

Symbo	Dimension in mm		Dimension in inch	
	MIN.	MAX.	MIN.	MAX.
A	2.692	3.099	0.106	0.122
B	1.397	1.803	0.055	0.071
C	-----	1.450	-----	0.057
D	0.300	0.550	0.012	0.022
F	0.838	1.041	0.033	0.041
H	0.080	0.254	0.003	0.010
I	0.050	0.150	0.002	0.006
J	2.600	3.000	0.102	0.118
M	0.300	0.600	0.012	0.024
θ	0	10°	0°	10°

SOT-23-8L:



VARIATION (ALL DIMENSIONS SHOWN IN MM)

SYMBOL	MIN.	NOM.	MAX.
A	-	-	1.45
A1	0.00	-	0.15
A2	0.90	1.15	1.30
b	0.22	-	0.38
c	0.08	-	0.22
D	2.90 BSC.		
E	2.80 BSC.		
E1	1.60 BSC.		
e	0.65 BSC.		
e1	1.95 BSC.		
L	0.30	0.45	0.60
L1	0.60 REF.		
L2	0.25 BSC.		
R	0.10	-	-
R1	0.10	-	0.25
θ	0°	4°	8°
$\theta 1$	5°	10°	15°

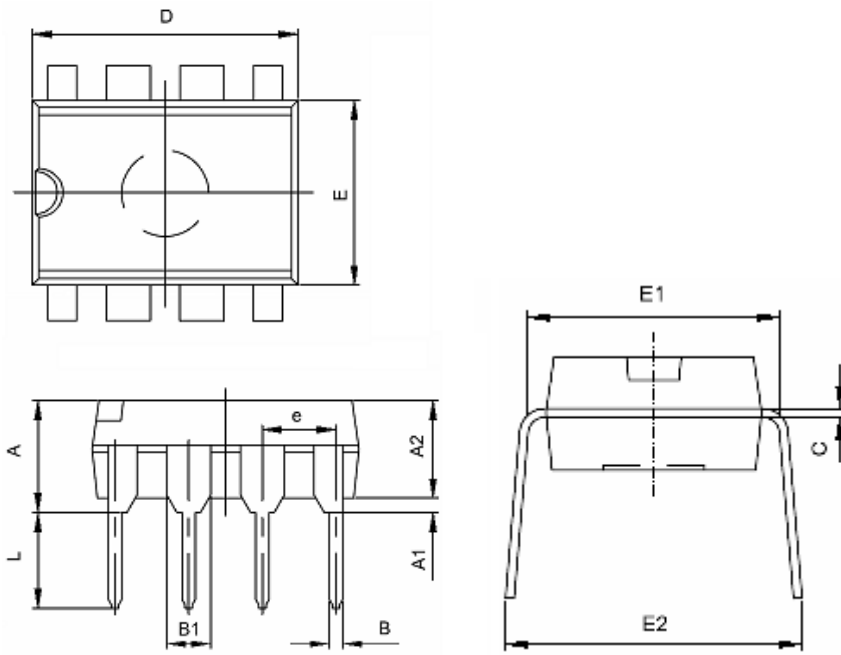
NOTE :
1. JEDEC OUTLINE : MO-178 BA.

EST.2700X Green-Mode PWM Flyback (SSR) Controller



EST.2700X Green-Mode PWM Flyback (SSR) Controller

DIP-8 Package



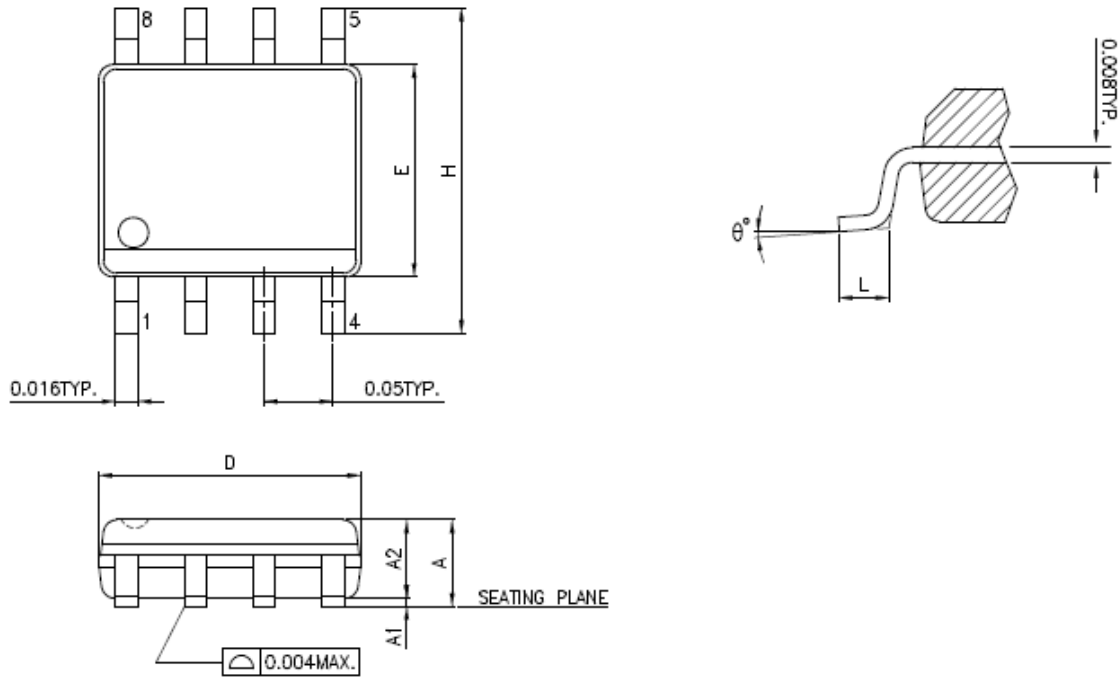
Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	3.710	4.310	0.146	0.170
A1	0.510		0.020	
A2	3.200	3.600	0.126	0.142
B	0.360	0.560	0.014	0.022
B1	1.524(TYP)		0.060(TYP)	
C	0.204	0.360	0.008	0.014
D	9.000	9.400	0.354	0.370
E	6.200	6.600	0.244	0.260
E1	7.620(TYP)		0.300(TYP)	
e	2.540(TYP)		0.100(TYP)	
L	3.000	3.600	0.118	0.142
E2	8.200	9.400	0.323	0.370

EST.2700X Green-Mode PWM Flyback (SSR) Controller



EST.2700X Green-Mode PWM Flyback (SSR) Controller

SOP-8 Package (mm)

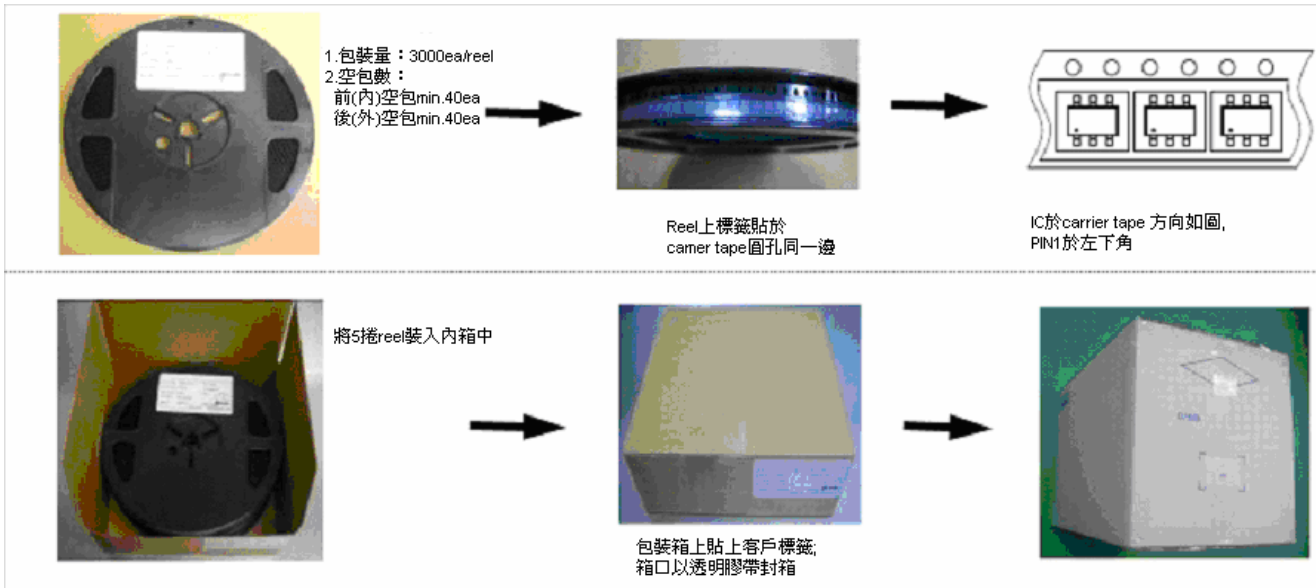


Symbols	Dimensions In Inches			Dimensions In millimeters		
	MIN.	NOR.	MAX.	MIN.	NOR.	MAX.
A	0.050	0.061	0.072	1.270	1.549	1.829
A1	0.000	-----	0.010	0.000	-----	0.254
A2	-----	-----	0.062	-----	-----	1.575
D	0.185	0.193	0.200	4.699	4.902	5.080
E	0.147	0.154	0.160	3.734	3.912	4.064
H	0.225	0.237	0.249	5.715	6.020	6.325
L	0.013	0.033	0.053	0.330	0.838	1.346
θ	0°	4°	8°	0°	4°	8°

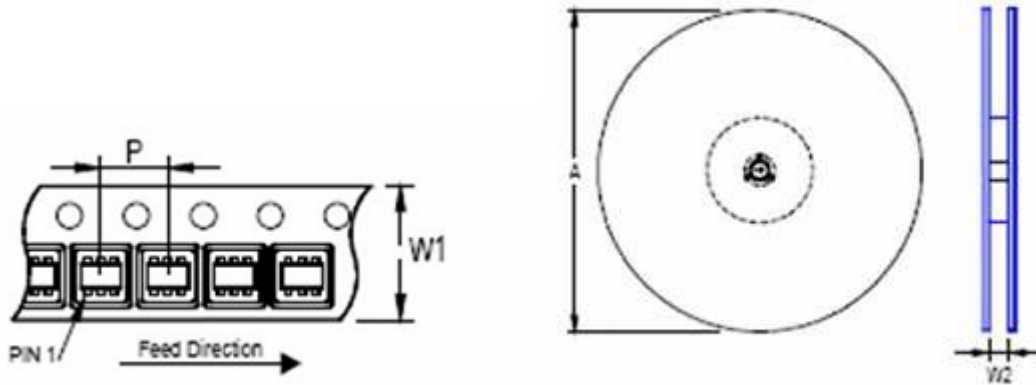
EST.2700X Green-Mode PWM Flyback (SSR) Controller



Shipping packing



Tape Reel Data



Package Type SOT-26	Tape Size (W1) (mm)	Pocket Pitch (P) (mm)	Reel Size (A) (mm)	Reel Width (W2) Min./Max. (mm)	Units Per Reel pcs.
6 Lead	8	4	180	8.4/9.9	3000

EST.2700X

Green-Mode PWM Flyback (SSR) Controller



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